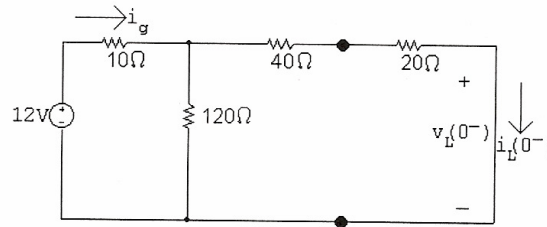


Homework 6 Solutions

1)

[a] $i_o(0^-) = 0$ since the switch is open for $t < 0$.

[b] For $t = 0^-$ the circuit is:

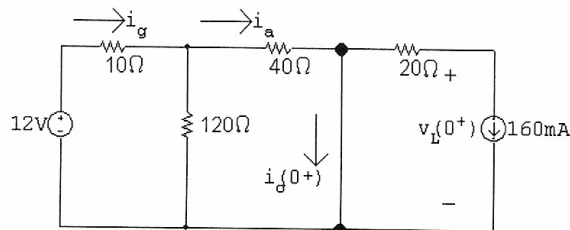


$$120\Omega \parallel 60\Omega = 40\Omega$$

$$\therefore i_g = \frac{12}{10 + 40} = 0.24\text{ A} = 240\text{ mA}$$

$$i_L(0^-) = \left(\frac{120}{180}\right) i_g = 160\text{ mA}$$

[c] For $t = 0^+$ the circuit is:



$$120\Omega \parallel 40\Omega = 30\Omega$$

$$\therefore i_g = \frac{12}{10 + 30} = 0.30\text{ A} = 300\text{ mA}$$

$$i_a = \left(\frac{120}{160}\right) 300 = 225\text{ mA}$$

$$\therefore i_o(0^+) = 225 - 160 = 65\text{ mA}$$

[d] $i_L(0^+) = i_L(0^-) = 160\text{ mA}$

[e] $i_o(\infty) = i_a = 225\text{ mA}$

[f] $i_L(\infty) = 0$, since the switch short circuits the branch containing the 20Ω resistor and the 100 mH inductor.

$$[g] \tau = \frac{L}{R} = \frac{100 \times 10^{-3}}{20} = 5\text{ ms}; \quad \frac{1}{\tau} = 200$$

$$\therefore i_L = 0 + (160 - 0)e^{-200t} = 160e^{-200t}\text{ mA}, \quad t \geq 0$$

[h] $v_L(0^-) = 0$ since for $t < 0$ the current in the inductor is constant

[i] Refer to the circuit at $t = 0^+$ and note:

$$20(0.16) + v_L(0^+) = 0; \quad \therefore v_L(0^+) = -3.2\text{ V}$$

[j] $v_L(\infty) = 0$, since the current in the inductor is a constant at $t = \infty$.

$$[k] v_L(t) = 0 + (-3.2 - 0)e^{-200t} = -3.2e^{-200t}\text{ V}, \quad t \geq 0^+$$

$$[l] i_o = i_a - i_L = 225 - 160e^{-200t}\text{ mA}, \quad t \geq 0^+$$

Homework 6 Solutions

2)

$$w(0) = \frac{1}{2}(20 \times 10^{-3})(11.2)^2 = 1254.4 \text{ mJ}$$

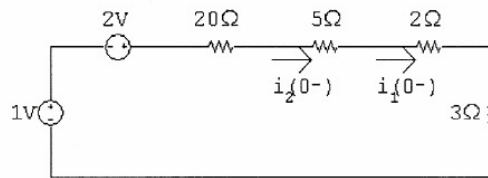
$$p_{30i_{\Delta}} = -30i_{\Delta}i_L = -30(-4.9e^{-2625t})(11.2e^{-2625t}) = 1646.4e^{-5250t} \text{ W}$$

$$w_{30i_{\Delta}} = \int_0^{\infty} 1646.4e^{-5250t} dt = 1646.4 \frac{e^{-5250t}}{-5250} \Big|_0^{\infty} = 313.6 \text{ mJ}$$

$$\% \text{ dissipated} = \frac{313.6}{1254.4}(100) = 25\%$$

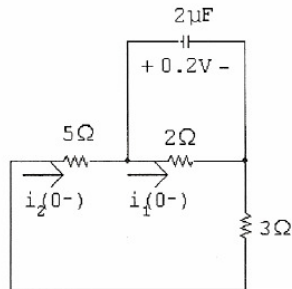
3)

[a] $t < 0$:



$$i_1(0^-) = i_2(0^-) = \frac{3 \text{ V}}{30 \Omega} = 100 \text{ mA}$$

[b] $t > 0$:



$$i_1(0^+) = \frac{0.2}{2} = 100 \text{ mA}$$

$$i_2(0^+) = \frac{-0.2}{8} = -25 \text{ mA}$$

[c] Capacitor voltage cannot change instantaneously, therefore,

$$i_1(0^-) = i_1(0^+) = 100 \text{ mA}$$

[d] Switching can cause an instantaneous change in the current in a resistive branch. In this circuit

$$i_2(0^-) = 100 \text{ mA} \quad \text{and} \quad i_2(0^+) = -25 \text{ mA}$$

[e] $v_c = 0.2e^{-t/\tau} \text{ V}, \quad t \geq 0 \quad R_e = 2 \parallel (5 + 3) = 1.6 \Omega$

$$\tau = 1.6(2 \times 10^{-6}) = 3.2 \times 10^{-6} \text{ s}$$

$$v_c = 0.2e^{-312,500t} \text{ V}, \quad t \geq 0$$

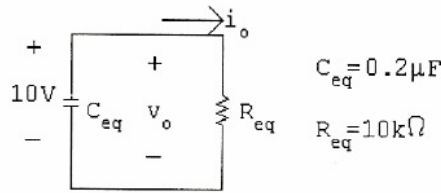
$$i_1 = \frac{v_c}{2} = 0.1e^{-312,500t} \text{ A}, \quad t \geq 0$$

[f] $i_2 = \frac{-v_c}{8} = -25e^{-312,500t} \text{ mA}, \quad t \geq 0^+$

Homework 6 Solutions

4)

[a] The equivalent circuit for $t > 0$:



$$\tau = 2 \text{ ms}; \quad 1/\tau = 500$$

$$v_o = 10e^{-500t} \text{ V}, \quad t \geq 0$$

$$i_o = e^{-500t} \text{ mA}, \quad t \geq 0^+$$

$$i_{24k\Omega} = e^{-500t} \left(\frac{16}{40} \right) = 0.4e^{-500t} \text{ mA}, \quad t \geq 0^+$$

$$p_{24k\Omega} = (0.16 \times 10^{-6} e^{-1000t})(24,000) = 3.84e^{-1000t} \text{ mW}$$

$$w_{24k\Omega} = \int_0^\infty 3.84 \times 10^{-3} e^{-1000t} dt = -3.84 \times 10^{-6}(0 - 1) = 3.84 \mu\text{J}$$

$$w(0) = \frac{1}{2}(0.25 \times 10^{-6})(40)^2 + \frac{1}{2}(1 \times 10^{-6})(50)^2 = 1.45 \text{ mJ}$$

$$\% \text{ diss } (24 \text{ k}\Omega) = \frac{3.84 \times 10^{-6}}{1.45 \times 10^{-3}} \times 100 = 0.26\%$$

[b] $p_{400\Omega} = 400(1 \times 10^{-3} e^{-500t})^2 = 0.4 \times 10^{-3} e^{-1000t}$

$$w_{400\Omega} = \int_0^\infty p_{400} dt = 0.40 \mu\text{J}$$

$$\% \text{ diss } (400\Omega) = \frac{0.4 \times 10^{-6}}{1.45 \times 10^{-3}} \times 100 = 0.03\%$$

$$i_{16k\Omega} = e^{-500t} \left(\frac{24}{40} \right) = 0.6e^{-500t} \text{ mA}, \quad t \geq 0^+$$

$$p_{16k\Omega} = (0.6 \times 10^{-3} e^{-500t})^2 (16,000) = 5.76 \times 10^{-3} e^{-1000t} \text{ W}$$

$$w_{16k\Omega} = \int_0^\infty 5.76 \times 10^{-3} e^{-1000t} dt = 5.76 \mu\text{J}$$

$$\% \text{ diss } (16k\Omega) = 0.4\%$$

[c] $\sum w_{\text{diss}} = 3.84 + 5.76 + 0.4 = 10 \mu\text{J}$

$$w_{\text{trapped}} = w(0) - \sum w_{\text{diss}} = 1.45 \times 10^{-3} - 10 \times 10^{-6} = 1.44 \text{ mJ}$$

$$\% \text{ trapped} = \frac{1.44}{1.45} \times 100 = 99.31\%$$

Check: $0.26 + 0.03 + 0.4 + 99.31 = 100\%$

Homework 6 Solutions

5)

Summing currents leaving the inverting input

$$\frac{v_N(t)}{R} + C \cdot \frac{d}{dt}(v_N(t) - v_O(t)) = 0$$

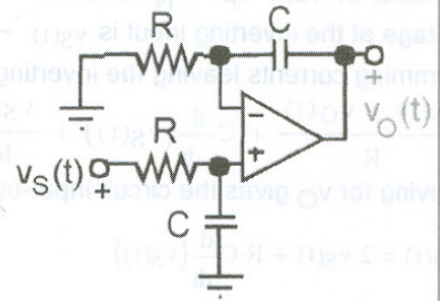
Solving for v_O gives

$$R \cdot C \cdot \frac{d}{dt} v_O(t) = v_N(t) + R \cdot C \cdot \frac{d}{dt}(v_N(t))$$

Summing currents leaving the noninverting input

$$\frac{v_P(t) - v_S(t)}{R} + C \cdot \frac{d}{dt}(v_P(t)) = 0$$

Solving for v_S gives $v_S(t) = v_P(t) + R \cdot C \cdot \frac{d}{dt}(v_P(t))$

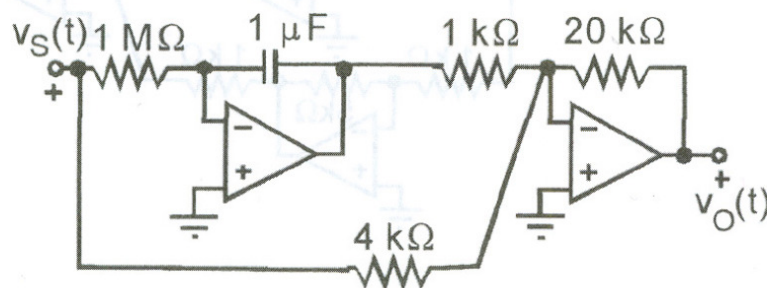


For an ideal OP AMP $v_P = v_N$ and as a result $v_P(t) + R \cdot C \cdot \frac{d}{dt}(v_P(t)) = v_N(t) + R \cdot C \cdot \frac{d}{dt}(v_N(t))$ which means

that $v_S(t) = R \cdot C \cdot \frac{d}{dt} v_O(t)$ integrating both sides gives $v_O(t) = \frac{1}{R \cdot C} \int_0^t v_S(x) dx + v_O(0)$ QED

6) The required input-output relationship is $v_O(t) = -5 \cdot v_S(t) + 20 \cdot \int_0^t v_S(x) dx$.

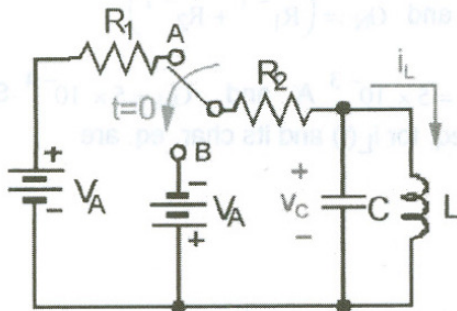
One of many possible designs is shown below



Homework 6 Solutions

7) Parallel RLC circuit with: $R_1 := 20000 = 20 \text{ k}\Omega$ $R_2 := 4000 = 4 \text{ k}\Omega$ $L := 1.6 \text{ H}$

$$C := 1.25 \cdot 10^{-6} = 1.25 \text{ }\mu\text{F} \quad V_A := 24 \text{ V}$$



For $t < 0$ sw in Pos A: $v_C(0) = 0$ and

$$i_L(0) = \frac{V_A}{R_1 + R_2} = 1 \cdot 10^{-3} \text{ for } t > 0 \text{ sw in Pos. B:}$$

$$\text{The Norton equivalent circuit is } i_N = \frac{-V_A}{R_2} = -6 \cdot 10^{-3} \text{ A}$$

$$\text{and } G_N := (R_2)^{-1} \text{ hence } G_N = 2.5 \times 10^{-4} \text{ S.}$$

The circuit differential eq. and its characteristic eq. are

$$L \cdot C \cdot \frac{d^2}{dt^2} i_L(t) + G_N \cdot L \cdot \frac{d}{dt} i_L(t) + i_L(t) = i_N \text{ and}$$

$$L \cdot C \cdot s^2 + G_N \cdot L \cdot s + 1 = 0 \text{ whose roots are}$$

$$\text{polyroots} \left(\begin{pmatrix} 1 & G_N \cdot L & L \cdot C \end{pmatrix}^T \right) = \begin{pmatrix} -100 - 700j \\ -100 + 700j \end{pmatrix} \quad \text{The circuit is underdamped so the natural response}$$

$$\text{has the form } i_{LN}(t) = e^{-100 \cdot t} \cdot (K_1 \cdot \cos(700t) + K_2 \cdot \sin(700t)).$$

$$\text{The forced response is } i_{LF} = i_N = -6 \cdot 10^{-3}.$$

$$\text{The total response is } i_L(t) = i_{LN}(t) + i_{LF}(t) = e^{-100 \cdot t} \cdot (K_1 \cdot \cos(700t) + K_2 \cdot \sin(700t)) + (-6 \cdot 10^{-3}).$$

$$\text{The initial condition } i_L(0) = 1 \cdot 10^{-3} \text{ implies } K_1 - 6 \times 10^{-3} = 1 \cdot 10^{-3} \text{ which means } K_1 = 7 \cdot 10^{-3} \text{ A.}$$

$$\text{The initial condition } \frac{d}{dt} i_L(0) = \frac{1}{L} \cdot v_C(0) = 0 \text{ implies } -100 \cdot K_1 + 700 \cdot K_2 = 0 \text{ which means}$$

$$K_2 = \frac{K_1}{7} = 10^{-3} \text{ A.}$$

$$\text{For } t > 0 \text{ the inductor current is } i_L(t) = 10^{-3} \cdot [e^{-100 \cdot t} \cdot (7 \cdot \cos(700t) + \sin(700t)) - 6] \text{ A.}$$

Using KVL and the inductor i-v characteristic leads to

$$v_C(t) = v_L(t) = L \cdot \frac{d}{dt} i_L(t) = 1.6 \cdot \frac{d}{dt} [10^{-3} \cdot e^{-100 \cdot t} \cdot (7 \cdot \cos(700t) + \sin(700t)) - 6] \text{ which reduces to}$$

$$v_C(t) = -8 \cdot e^{-100 \cdot t} \cdot \sin(700t) \text{ V}$$

Homework 6 Solutions

8)

When the sw is in position A the capacitor discharges through the short and the inductor current is zero, that is $v_C(0) = 0$ and $i_L(0) = 0$. At $t = 0$ the sw moves to position B where the capacitor and 10-k Ω resistor form a 1st order RC circuit whose charging time constant is

$R \cdot C = 10^4 \cdot 10^{-6} = 0.01$. If left in this position for a long time the final value of the capacitor voltage would be $v_C = 10$ V. Thus, for $0 \leq t \leq 10$ ms the capacitor voltage is $v_C(t) = 10(1 - \exp(-100t))$ V. The capacitor voltage at $t = 10$ ms is $v_C(0.01) = 10(1 - \exp(-1)) = 6.32$ V. At $t = 10$ ms the sw is moved to position C where the capacitor and 1-H inductor form an undamped ($\zeta = 0$) 2nd order system whose undamped natural frequency is $\omega_0 = \frac{1}{\sqrt{L \cdot C}} = 1000$ rad/s. Thus, for $t > 10$ ms the capacitor voltage is an undamped sinusoidal oscillation with a frequency of 1000 rad/s and amplitude of $v_C(0.01) = 6.321$ V.

(a) If R is reduced to 1 k Ω the charging time constant decreases to $R \cdot C = 0.001$, the capacitor voltage at $t = 10$ ms becomes $v_C(0.01) = 10 \cdot (1 - \exp(-10)) = 10$ V, and $\omega_0 = \frac{1}{\sqrt{L \cdot C}}$ is unchanged. Hence the

amplitude increases and the frequency remains the same.

(b) If L is reduced to 100 mH the charging time constant is unchanged, but the undamped natural frequency becomes $\omega_0 = \frac{1}{\sqrt{L \cdot C}} = 3162$ rad/s. Hence the amplitude is unchanged and the frequency increases.

