

ECE 2C Mid-Term Exam A Solutions

May 8, 2012

Do not open exam until instructed to.

Closed book: Crib sheet and 1 page personal notes permitted

There are 2 problems on this exam, and you have 75 minutes.

Use any and all reasonable approximations (5% accuracy is fine.), ***AFTER STATING and approximately Justifying them.***

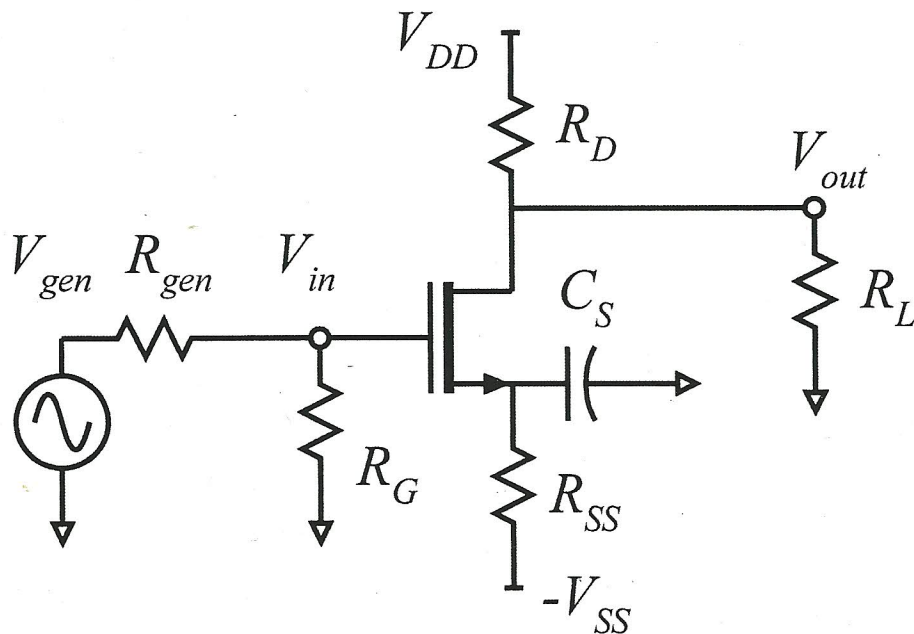
Name: _____

Problem	Points Received	Points Possible
1a		10
1b		4
1c		10
1d		15
1e		5
1f		10
1g		10
2a		15
2b		15
2c		5
total		100

Problem 1, 65 points

Transistor circuit analysis

You will be working on the circuit below:



Q1 is a mobility-limited FET, ie. $I_d = (\mu C_{ox} W_g / 2L_g)(V_{gs} - V_{th})^2(1 + \lambda V_{ds})$ where $(\mu C_{ox} W_g / 2L_g) = 4 \text{ mA/V}^2$, $\lambda = 0.1 \text{ V}^{-1}$, and $V_{th} = 0.5 \text{ V}$.

$V_{DD} = +2.0 \text{ Volts}$, $-V_{SS} = -2 \text{ Volts}$

C_S is very big and has negligible AC impedance.

$R_L = 5 \text{ k}\Omega$, $R_{gen} = 100 \text{ k}\Omega$, $R_G = 1 \text{ M}\Omega$

Part a, 10 points

DC bias.

Q1 is to be biased with 1 mA drain current.

The drain of Q1 is to be biased at 0.0 Volts

Ignore λ while solving this part

Find: $R_{SS} = \underline{1\text{ k}\Omega}$ $R_D = \underline{2\text{ k}\Omega}$

We are given that Q1 is a mobility-limited FET with current equation

$$I_D = (\mu C_{ox} W_g / 2 L_g) (V_{GS} - V_{th})^2 (1 + \lambda V_{DS})$$

We plug in $\mu C_{ox} W_g / 2 L_g = 4\text{ mA}\cdot\text{V}^{-2}$, $V_{th} = 0.5\text{ V}$, and assume $\lambda = 0$ to get

$$(2) \quad V_{GS} = \left(\frac{I_D}{\mu C_{ox} W_g / 2 L_g} \right)^{1/2} + V_{th} = \left(\frac{10^{-3}\text{ A}}{4\text{ mA}\cdot\text{V}^{-2}} \right)^{1/2} + 0.5\text{ V} = 1\text{ V}$$

And now we can solve for R_{SS} , R_D :

Since V_{gen} is a sinusoidal signal, we assume that its average DC value is zero, and hence $V_g = 0$.

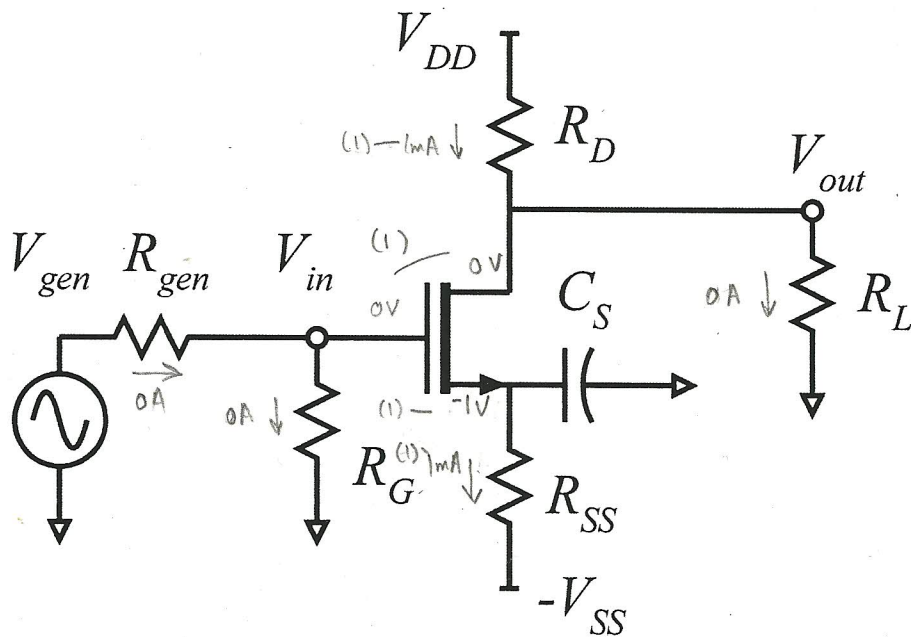
$$(2) \quad \text{Thus } V_g = -V_{GS} = -1\text{ V.}$$

$$(3) \quad R_{SS} = \frac{V_g - V_{SS}}{I_D} = \frac{(-1 - 2)\text{ V}}{10^{-3}\text{ A}} = 1\text{ k}\Omega$$

$$(3) \quad R_D = \frac{V_{DD} - V_D}{I_D} = \frac{(2 - 0)\text{ V}}{10^{-3}\text{ A}} = 2\text{ k}\Omega$$

Part b, 5 points

DC bias



On the circuit diagram above, label the DC voltages at **ALL nodes** and the DC currents through **ALL resistors**

Part c, 10 points

Find the small signal parameters of Q1. Use the mobility-limited model.

$$g_m = 4.4 \text{ mS}$$

$$R_{DS} = 11 \text{ k}\Omega$$

$$\begin{aligned} (3) - g_m &= \frac{\partial I_D}{\partial V_{GS}} = (\mu C_{ox} W/L_g) (V_{GS} - V_{th}) (1 + \lambda V_{DS}) \\ &= 2 (4 \text{ mA} \cdot \text{V}^{-2}) (1 - 0.5 \text{ V}) (1 + 0.1 \text{ V}^{-1} \cdot 1 \text{ V}) \\ &= 4.4 \text{ mS} - (2) \end{aligned}$$

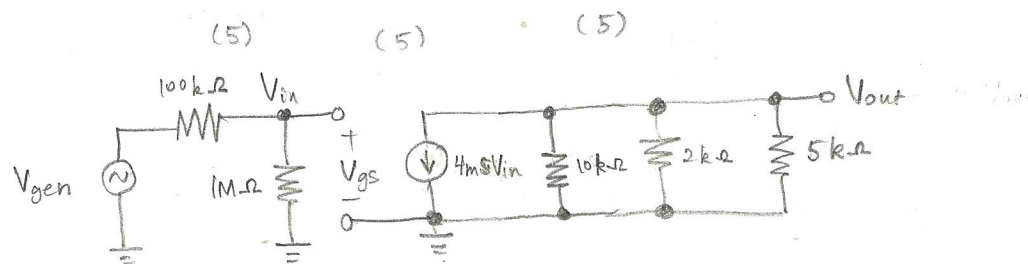
$$(3) - R_{DS} = \frac{\frac{1}{\lambda} + V_{DS}}{I_D} = \frac{10 \text{ V} + 0 \text{ V}}{1 \text{ mA}} = 10 \text{ k}\Omega - (2)$$

OR

$$\begin{aligned} (3) - R_{DS} &= \left(\frac{\partial I_D}{\partial V_{DS}} \right)^{-1} = (\mu C_{ox} W/L_g) (V_{GS} - V_{th})^2 \lambda \\ &= (4 \text{ mA} \cdot \text{V}^{-2}) (1 - 0.5 \text{ V})^2 (0.1 \text{ V}^{-1}) \\ &= 10 \text{ k}\Omega - (2) \end{aligned}$$

Part d, 15 points

Replacing the transistor with its small-signal model, draw a small-signal equivalent circuit diagram for the amplifier. Give values for all elements on the diagram.



Part e, 5 points.

Find the small signal voltage gain (V_{out}/V_{in}) of Q1.

$$V_{out}/V_{in} = \underline{-5.54 \text{ V/V}}$$

$$R_{LEQ} = R_{DS} \parallel R_L \parallel R_D = \frac{2k\Omega \cdot 5k\Omega \cdot 11k\Omega}{2k\Omega \cdot 5k\Omega + 2k\Omega \cdot 11k\Omega + 5k\Omega \cdot 11k\Omega}$$

$$R_{LEQ} \approx 1.26 k\Omega$$

$$\frac{V_{out}}{V_{in}} = \frac{-g_m V_{gs} R_{LEQ}}{V_{in}}, \quad V_{in} = V_{gs}$$

$$\frac{V_{out}}{V_{in}} = -g_m R_{LEQ} = -4.4mS \cdot 1.26 k\Omega \approx -5.54$$

Part f, 10 points

Find the *** amplifier *** input resistance, V_{in}/V_{gen} , and V_{out}/V_{gen}

$$R_{in, amplifier} = \underline{1 \text{ M}\Omega}$$

$$V_{in}/V_{gen} = \underline{0.90 \text{ V/V}}$$

$$(V_{out}/V_{gen}) = \underline{-5.04 \text{ V/V}}$$

$$R_{in, amplifier} = \frac{V_{in}}{I_{in}} = R_g \quad \text{No current into gate}$$
$$= 1 \text{ M}\Omega$$

$$\frac{V_{in}}{V_{gen}} = \frac{\frac{V_{gen} R_g}{R_g + R_{gen}}}{V_{gen}} = \frac{R_g}{R_g + R_{gen}} = \frac{1 \text{ M}\Omega}{1 \text{ M}\Omega + 100 \text{ k}\Omega}$$

$$\frac{V_{in}}{V_{gen}} = 0.90$$

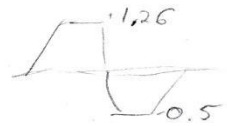
$$\frac{V_{out}}{V_{gen}} = \frac{V_{out}}{V_{in}} \times \frac{V_{in}}{V_{gen}} = -5.54 \times 0.90 \approx -5.04$$

Part g, 10 points

Now you must find the maximum signal swings. Find the output voltage due to the knee voltage and due to cutoff in Q1.

Cutoff of Q1; Maximum ΔV_{out} resulting = + 1.26 V

Knee voltage of Q1; Maximum ΔV_{out} resulting = - 0.5 V



$$\Delta V_{out} \uparrow_{max} = R_{LEQ} \cdot \Delta I_{D \downarrow max} = 1.26 k \times 1 mA = 1.26 V$$

$\Delta I_{D \downarrow max}$ limited by the 1mA DC bias current

$$V_{dg} = -V_t \quad \text{at the knee}$$

$$V_{DS, min} = V_{gs} + V_{dg} = 1 - 0.5 = 0.5 V$$

$$\Delta V_{out} \downarrow_{max} = V_{DS, bias} - V_{DS, min} = 1 V - 0.5 V = 0.5 V$$

Problem 2, 35 points

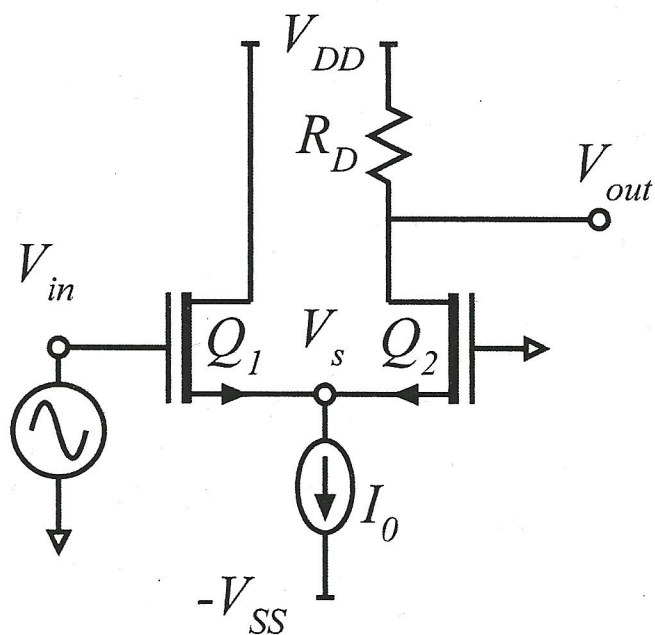
Small signal analysis, Nodal analysis

Part a, 15 points

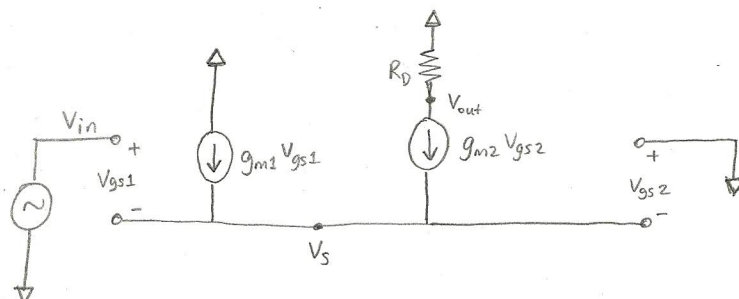
In the circuit at right, the FETs both have $g_{m1} = g_{m2} = g_m$ and $G_{ds1} = G_{ds2} = 0$.

V_{DD} and $-V_{SS}$ are DC bias voltages, while I_0 is a DC bias current source.

Replacing the two FETs with their small-signal models, draw below a small-signal diagram of the circuit. Be sure to label the nodes V_{in} , V_{out} , and V_s in the diagram.



- Replace FETs with small signal models. (5 pts)
- Replace DC voltage sources with short circuits. (3 pts)
- Replace DC current source with open circuit. (3 pts)
- Correct connections and labels. (4 pts)



* Since $G_{ds1} = G_{ds2} = 0$,
 $R_{ds1} = R_{ds2} = \infty$,
 we can ignore.

Part b, 15 points

Using nodal analysis, determine an algebraic expression for V_{out}/V_{in}

Nodal equation at V_s : $\underline{g_m(V_{in} - V_s) - g_m V_s = 0}$

Nodal equation at V_{out} : $\underline{\frac{V_{out}}{R_D} - g_m V_s = 0}$

$V_{out}/V_{in} = \underline{\frac{1}{2} g_m R_D}$

At V_s : $g_{m1} V_{gs1} + g_{m2} V_{gs2} = 0$
 $g_m(V_{in} - V_s) + g_m(0 - V_s) = 0$
 $g_m V_{in} = 2g_m V_s$
 $V_{in} = 2V_s$

At V_{out} : $(V_{out} - 0)\frac{1}{R_D} + g_{m2} V_{gs2} = 0$
 $\frac{V_{out}}{R_D} + g_m(0 - V_s) = 0$
 $V_{out} = g_m R_D V_s$

$\frac{V_{out}}{V_{in}} = \frac{g_m R_D V_s}{2V_s} = \underline{\underline{\frac{1}{2} g_m R_D}}$

Σ currents at node = 0

- Correct nodal equations for answer from part (a). (8 pts)
- Solving for $\frac{V_{out}}{V_{in}}$. (7 pts)

Part C, 5 points

$$R_D = 2 \text{ k}\Omega, g_m = 10 \text{ mS}.$$

Find the numerical value of V_{out}/V_{in} :

$$V_{out}/V_{in} \quad \underline{\quad 10 \quad}$$

- Plug in numbers into answer from part (b) correctly. (5 pts)

$$\frac{V_{out}}{V_{in}} = \frac{1}{2} g_m R_D = \frac{1}{2} (2 \text{ k}\Omega)(10 \text{ mS}) = \underline{\underline{10}}$$