

ECE 137 A Mid-Term Exam

February 7, 2007

Do not open exam until instructed to.

Closed book: Crib sheet and 1 page personal notes permitted

There are 2 problems on this exam, and you have 50 minutes.

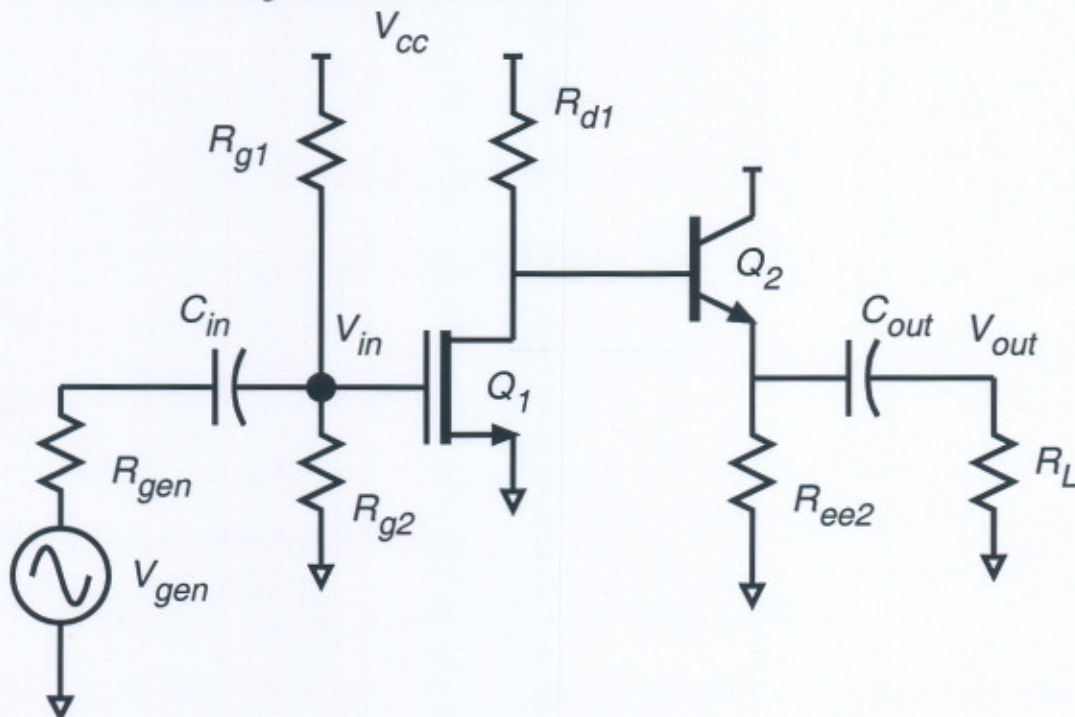
Use any and all reasonable approximations (5% accuracy is fine.), ***AFTER STATING and approximately Justifying them.***

Name: Solomon

Problem	Points Received	Points Possible
1a		7
1b		7
1c		6
1d		15
1e		15
1f		6
1g		14
2a		12
2b		13
2c		5
total		100

Problem 1, 70 points

You will be working on the circuit below:



Q1: $V_{th}=0.3\text{ V}$, $v_{sat}c_{ox}W_g = 10\text{ mA/V}$, $\lambda = 0\text{ 1/Volts}$

Q2: $\beta=100$, $V_A=\text{infinity V}$

+Vcc= +3.0 volts. $R_{gen}=100\text{ kOhm}$ $R_L=1\text{ kOhm}$

$R_{ee2}=????$ $R_{d1}=???$ $R_{g1}=????$

$R_{g2}= 400\text{ kOhm}$

C_{in1} and C_{out} are very big and have negligible AC impedance.

Part a, 7 points

DC bias.

Q1 is to be biased with 1 mA drain current.

Q2 is to be biased with 1 mA emitter current.

The DC emitter voltage of Q2 is +2.0 Volts.

Find: $R_{EE2} = 2k\Omega$ $R_{D1} = 297\Omega$ $R_{G1} = 2.6M\Omega$

1pt $R_{EE2} = 2V / 1mA = 2k\Omega$

1pt $V_{B2} = 2V + \phi = 2.7V$

2pt $R_{D1} = (V_{CC} - V_{B2}) / (I_{D1} + I_{B2})$

$I_{B2} = 1mA / \beta = 10\mu A$

$= (3V - 2.7V) / (1mA + 10\mu A) = 297\Omega$

1pt $V_{GS1} = ?$

wall

$V_{GS1} = V_{th} + I_D / g_m = 0.3V + 1mA / (60\mu A/V)$

$= 0.4V$

2pt. current in R_{G1}, R_{G2} : $I = 0.4V / R_{G2} = 1\mu A$

$R_{G2} = (3V - 0.4V) / 1\mu A = 2.6V / 1\mu A = 2.6M\Omega$

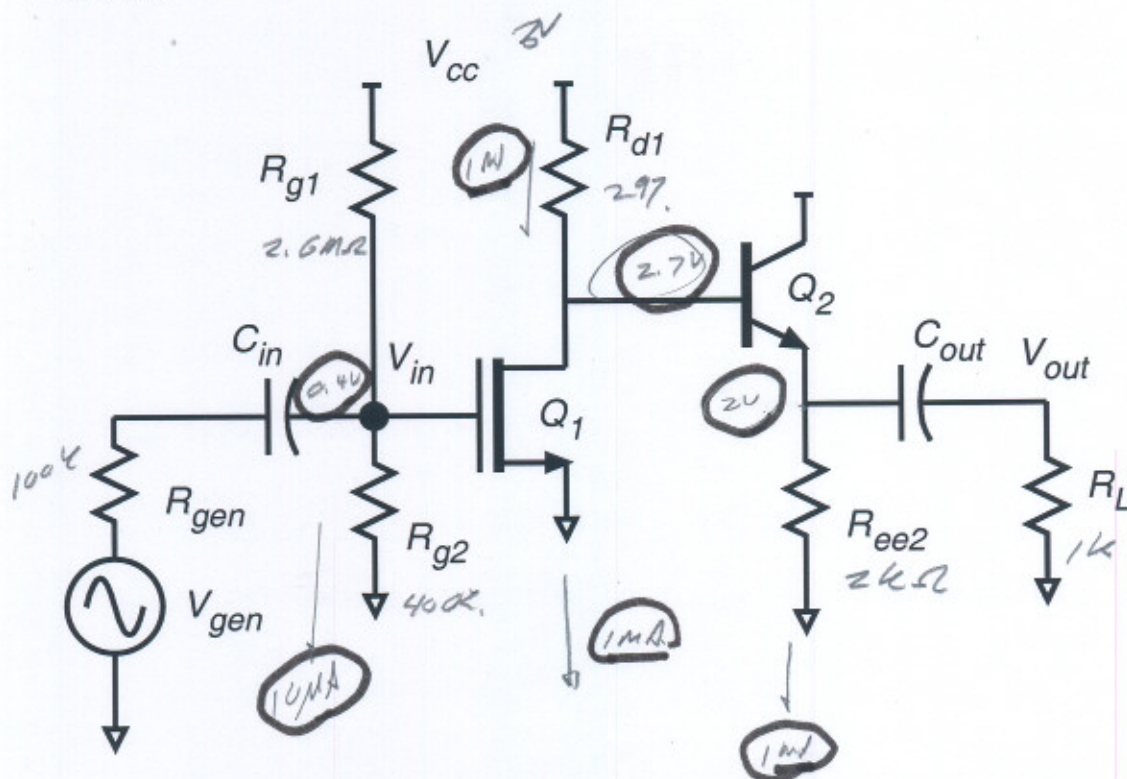
Part b, 7 points

DC bias

$$V_{d1} = 0.3V, g_m = 10 \mu A/V$$

$$\lambda = 0$$

$$\beta = 100, V_A = \infty$$



On the circuit diagram above, label the DC voltages at **ALL** nodes and the DC currents through **ALL** resistors

$$Q2: g_m = \frac{I}{V_T} = \frac{1}{26mV}$$

$$Q1: g_m = 10 \mu A/V$$

Part c, 6 points

Find the small signal parameters of Q1 and Q2.

Transistor Q1: $g_m = \underline{10 \text{ mS}}$ $R_{ds} = \underline{\infty}$

Transistor Q2: $g_m = \underline{38.5 \text{ mS}}$ $R_{ce} = \underline{\infty}$ $R_{be} = \underline{2.6 \text{ k}\Omega}$

$$Q2: g_m = \frac{1}{r_e} = \frac{I_E}{V_T} = \frac{1.2 \text{ A}}{26 \text{ mV}} = 38.5 \text{ mS} \quad]^2$$

$$R_{BE} = \beta r_e = 100 \cdot 26 \Omega = 2.6 \text{ k}\Omega \quad]^1$$

$$R_{CE} = \frac{V_{CE} - V_A}{I_C} = \frac{\infty}{I_C} = \infty \quad]^1$$

$$Q1: g_m = \sqrt{2 \mu_n C_{ox} W/L} I_D = 10 \text{ mS} \quad]^1$$

$$R_{ds} = \frac{1}{\lambda I_D} = \infty \quad]^1$$

Part d, 15 points.

Find the small signal voltage gain (V_{e2}/V_{b2}) of Q2 and Q2's small-signal input resistance.

$V_{e2}/V_{b2} = \underline{0.962}$

$R_{in,q2} = \underline{69.34k\Omega}$

Q2: EF

5 $\left[R_{eq} = 24 \parallel 1k = 667\Omega \right]$

$r_e = 26\Omega$

5 $\left[A_v = \frac{R_{eq}}{R_{eq} + r_e} = \frac{667}{667 + 26} = 0.962 \right]$

5 $\left[R_{i2} = \beta(r_e + R_{eq}) = 100(26 + 667) = 69.34k\Omega \right]$

Part e, 15 points

Find the small signal voltage gain (V_{d1}/V_{g1}) of Q1 and the *** amplifier *** input resistance.

$$V_{d1}/V_{g1} = \underline{\quad -2.96 \quad}$$

$$R_{in, \text{amplifier}} = \underline{\quad 347 \text{ k}\Omega \quad}$$

Q1: CS:

$$s \left[R_{eq} = R_{d1} \parallel R_{i2} = 297 \parallel 69.7 \text{ k}\Omega \right. \\ \left. \approx 298 \Omega \right]$$

$$s \left[A_v = -g_m R_{eq} = -10 \text{ mS} \cdot 298 \Omega = -2.96 \right]$$

$$s \left[R_{iQ1} = \infty \right. \\ \left. R_{in} = R_{g1} \parallel R_{g2} = 2.6 \text{ M}\Omega \parallel 400 \text{ k}\Omega \right. \\ \left. = 347 \text{ k}\Omega \right]$$

Part f, 6 points

Find (V_{out}/V_{in}) , (V_{in}/V_{gen}) and (V_{out}/V_{gen})

$$(V_{out}/V_{in}) = \underline{-2.85}$$

$$(V_{in}/V_{gen}) = \underline{0.776}$$

$$(V_{out}/V_{gen}) = \underline{-2.21}$$

$$2 \quad \left[\frac{V_o}{V_i} = A_{v1} A_{v2} = 0.962 \cdot (-2.96) = -2.85 \right]$$

$$2 \quad \left[\frac{V_i}{V_{gen}} = \frac{R_{in}}{R_{in} + R_{gen}} = \frac{347k\Omega}{347k\Omega + 100k\Omega} = 0.776 \right]$$

$$2 \quad \left[\frac{V_o}{V_{gen}} = \frac{V_o}{V_i} \cdot \frac{V_i}{V_{gen}} = (-2.85) \cdot (0.776) = -2.21 \right]$$

Part g, 14 points

Now you must find the maximum signal swings. Find the output voltage due to saturation and cutoff in Q2. *To save time, we are going to ignore the saturation and the knee voltage cutoff of Q1, even though it might be important. Give the sign (+ or -) in your answers below.*

Cutoff of Q2; Maximum ΔV_{out} resulting = -667 mV

Saturation of Q2; Maximum ΔV_{out} resulting = +500 mV

Q2 cutoff

Bias: $I_{CQ1, Q2} = 1 \text{ mA}$

mid $I_{C1} = 0 \text{ A}$

3 [$\Delta I_C = 1 \text{ mA max (decrease)}$

4 [$\Delta V_{out} = 1 \text{ mA} \cdot R_{eq} = 1 \text{ mA} \cdot 667 \Omega$
 $= 667 \text{ mV}$ negative

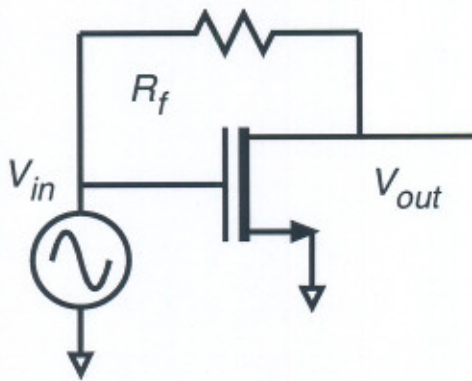
Q2 saturation

2 [Bias: $V_{CE} = 1 \text{ V}$

1 [Sat $V_{CE} = V_{CEsat} = 1/2 \text{ V}$

4 [$\Delta V_{CE} = \Delta V_{out} = 1/2 \text{ V}$ positive

Problem 2, 30 points
nodal analysis



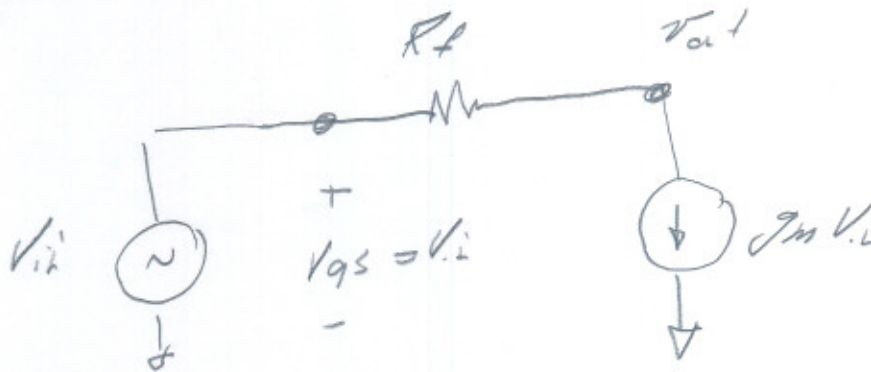
You will be working on the circuit to the left.

Ignore DC bias analysis. You don't need it.

The transistor has transconductance g_m . Its drain-source resistance R_{ds} is infinity (so you don't need to draw it!)

Part a, 12 points

Draw the small-signal equivalent circuit



Hard to give partial credit

Part b, 13 points

Find, by nodal analysis, a small-signal expression for V_{out}/V_{in} .

$V_{out}/V_{in} =$ _____

$\sum I = 0 @ V_{out}$

6 $\left[\frac{V_{out} - V_{in}}{R_f} + g_m V_{in} = 0 \right]$

$$V_{out} \left(\frac{1}{R_f} \right) = V_{in} \left[\frac{1}{R_f} - g_m \right]$$

7 $\left[\frac{V_{out}}{V_{in}} = 1 - g_m R_f \right]$

Part c, 5 points

$g_m = 10 \text{ mS}$ $R_f = 1000 \text{ Ohms}$

Give a numerical value for V_{out}/V_{in} .

$V_{out}/V_{in} = \underline{-9}$

3

$$\left[\begin{aligned} \frac{V_{out}}{V_{in}} &= 1 - g_m R_f \\ &= 1 - 10 \text{ mS} \cdot 1000 \Omega \\ &= -9 \end{aligned} \right.$$