

ECE 137 A Mid-Term Exam

February 7, 2007

Do not open exam until instructed to.

Closed book: Crib sheet and 1 page personal notes permitted

There are 2 problems on this exam, and you have 50 minutes.

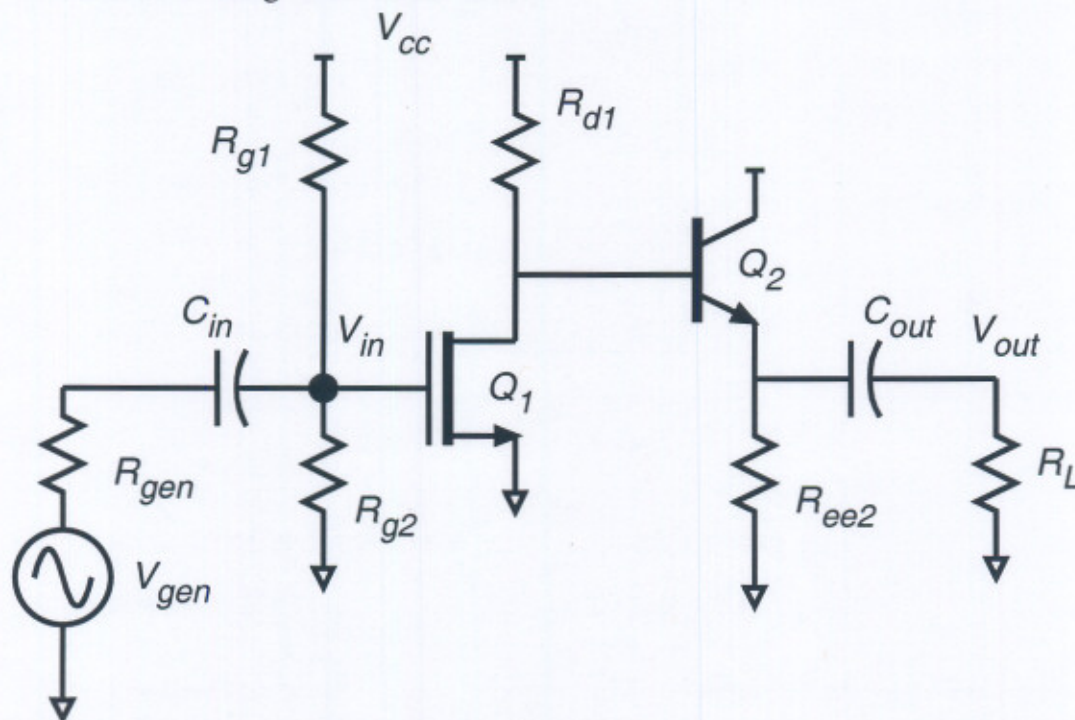
Use any and all reasonable approximations (5% accuracy is fine.), ***AFTER STATING and approximately Justifying them.***

Name: Soltani

Problem	Points Received	Points Possible
1a		7
1b		7
1c		6
1d		15
1e		15
1f		6
1g		14
2a		12
2b		13
2c		5
total		100

Problem 1, 70 points

You will be working on the circuit below:



Q1: $V_{th}=0.3$ V, $v_{sat}C_{ox}W_g = 10$ mA/V, $\lambda = 0$ 1/Volts

Q2 : $\beta=200$, $V_A = \text{infinity}$ V

+Vcc= +3.0 volts. $R_{gen}=100$ kOhm $R_L=1$ kOhm

$R_{ee2}=?????$ $R_{d1}=???$ $R_{g1}=????$

$R_{g2}= 500$ kOhm

Cin1 and Cout are very big and have negligible AC impedance.

Part a, 7 points

DC bias.

Q1 is to be biased with 1 mA drain current.

Q2 is to be biased with 2 mA emitter current.

The DC emitter voltage of Q2 is +1.3 Volts.

Find: $R_{EE2} = \underline{650 \Omega}$ $R_{D1} = \underline{990 \Omega}$ $R_{G1} = \underline{3.25 M\Omega}$

① $R_{EE2} = 1.3V / 2mA = 650 \Omega$

① $V_{B2} = 1.3V + \phi = 2V$

② $I_{B2} = 2mA / \beta = 2mA / 200 = 10\mu A$

② $R_{D1} = \frac{(3V - V_{E2})}{I_{D1} + I_{B1}} = \frac{3V - 2V}{1mA + 10\mu A} = 990 \Omega$

① $R_{G1} = ?$

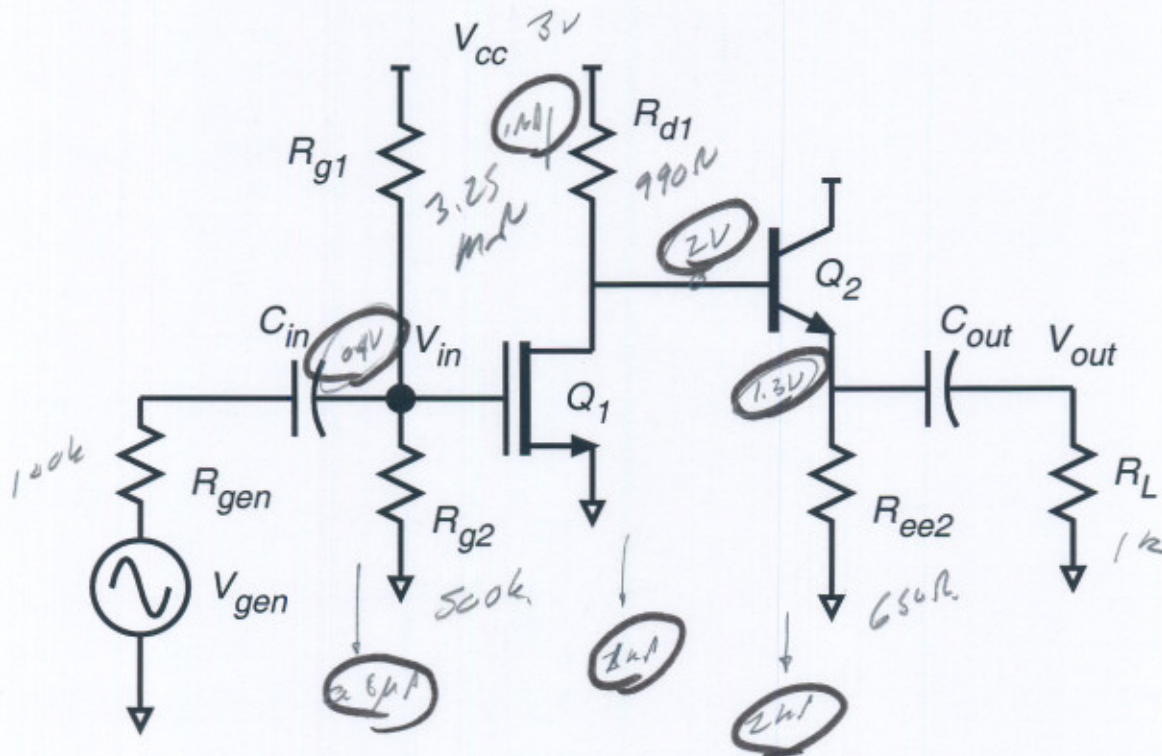
$V_{G1} = V_{B1} + I_D / \beta = 0.3V + 1mA / 10k = 0.4V$

② current i $R_{G1}, R_{G2} = V_{G1} / R_{G2} = 0.4V / 0.5M\Omega = 0.8\mu A$

$R_{G1} = \frac{3V - 0.4V}{0.8\mu A} = 3.25 M\Omega$

Q1: $V_{BE} = 0.3V$, $r_e = 10k\Omega$, $\lambda = 0$
Q2: $\beta = 200$, $V_A = \infty$

Q2: $\beta + 2\omega$, $V_A = V_C$



4

Part c, 6 points

Find the small signal parameters of Q1 and Q2.

Transistor Q1: $g_m = 10 \text{ mS}$ $R_{ds} = \infty$

Transistor Q2: $g_m = 76.9 \text{ mS}$ $R_{ce} = \infty$ $R_{be} = 2.6 \text{ k}\Omega$

Q1: $g_m = v_{gs} / C_{ox} W L = 10 \text{ mS} \quad] 2$
 $R_{ds} = 1 / \lambda I_{d1} = \infty \quad] 1$

Q2: $g_m = 1 / r_e = I_E / V_T = 1 / 13 \Omega = 76.9 \text{ mS} \quad] 1$
 $R_{ce} = \frac{V_A + V_{CE}}{I_C} = \infty \quad] 1$
 $R_{be} = \beta r_e = 200 \cdot 13 \Omega = 2.6 \text{ k}\Omega \quad] 1$

Part d, 15 points.

Find the small signal voltage gain (V_{e2}/V_{b2}) of Q2 and Q2's small-signal input resistance.

$V_{e2}/V_{b2} =$ 0.968

$R_{in,q2} =$ 81.4 k Ω

Q2: CK

$$s \left[R_{eq} = R_{EE2} \parallel R_c = 680\Omega \parallel 1k\Omega \right. \\ \left. = 394\Omega \right]$$

$$s \left[A_v = \frac{R_{eq}}{R_{eq} + r_e} = \frac{394}{394 + 13} = 0.968 \right]$$

$$s \left[R_{i2} = \beta (R_{eq} + r_e) = 200(394\Omega + 13\Omega) \right. \\ \left. = 81.4 k\Omega \right]$$

Part e, 15 points

Find the small signal voltage gain (V_d/V_{g1}) of Q1 and the *** amplifier *** input resistance.

$$V_d/V_{g1} = \underline{-9.78}$$

$$R_{in, \text{amplifier}} = \underline{433 \text{ k}\Omega}$$

$$s \left[\begin{array}{l} Q1: R_{eq} = R_{d1} \parallel R_{i2} \\ = 970 \Omega \parallel 81 \text{ k}\Omega \\ = 978 \Omega \end{array} \right.$$

$$s \left[\begin{array}{l} A_v = -g_m R_{eq} = -10 \text{ mS} \cdot 978 \Omega \\ = -9.78 \end{array} \right.$$

$$s. \left[\begin{array}{l} R_{i2} \uparrow = \infty \\ R_{i2} \downarrow = R_{i2} \parallel R_{g2} = 500 \text{ k}\Omega \parallel 3.2 \text{ M}\Omega \\ = 433 \text{ k}\Omega \end{array} \right.$$

Part f, 6 points

Find (V_{out}/V_{in}) , (V_{in}/V_{gen}) and (V_{out}/V_{gen})

$$(V_{out}/V_{in}) = \underline{-9.46}$$

$$(V_{in}/V_{gen}) = \underline{0.81}$$

$$(V_{out}/V_{gen}) = \underline{-7.69}$$

$$2 \left[\begin{aligned} V_o/V_i &= A_{v1} \cdot A_{v2} = 0.964 \cdot (-9.78) \\ &= -9.46 \end{aligned} \right]$$

$$2 \left[V_i/V_{gen} = \frac{R_{in}}{R_{in} + R_{gen}} = \frac{433\Omega}{433\Omega + 100\Omega} = 0.81 \right]$$

$$2 \left[V_o/V_g = \frac{V_o}{V_i} \cdot \frac{V_i}{V_g} = (-9.46)(0.81) = -7.69 \right]$$

Part g, 14 points

Now you must find the maximum signal swings. Find the output voltage due to saturation and cutoff in Q2. *To save time, we are going to ignore the saturation and the knee voltage cutoff of Q1, even though it might be important. Give the sign (+ or -) in your answers below.*

Cutoff of Q2; Maximum ΔV_{out} resulting = - 788 mV

Saturation of Q2; Maximum ΔV_{out} resulting = + 500 mV.

Q2 cutoff

3 [$I_{CQ} = 2 \text{ mA}$
 $\Delta I_{Cmax} = 2 \text{ mA decrease}$
 $R_{eq} = 394 \Omega$

4 [$\Delta V_{out} = 394 \Omega \cdot 2 \text{ mA} = 788 \text{ mV}$ Negative

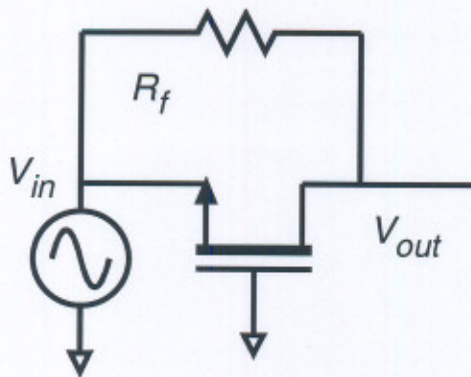
Q2 sat

2 [$V_{CE, BIAS} = 3V - 1.3V = 1.7V$

1 [$V_{CE sat} \approx 0.5V$

4 [$\Delta V_{out} = 1.7V - 0.5V = 1.2V$ positive

Problem 2, 30 points
nodal analysis



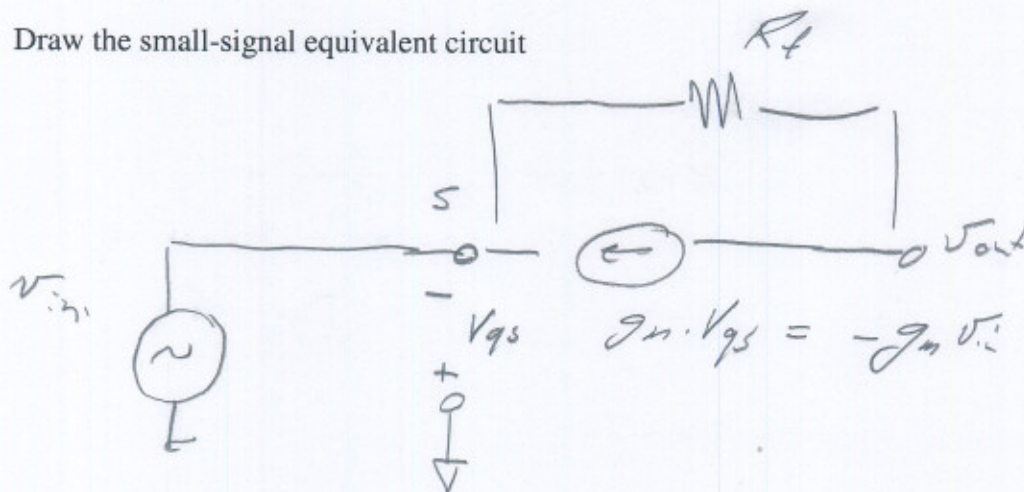
You will be working on the circuit to the left.

Ignore DC bias analysis. You don't need it.

The transistor has transconductance g_m . Its drain-source resistance R_{ds} is infinity (so you don't need to draw it!)

Part a, 12 points

Draw the small-signal equivalent circuit



Hard to give partial credit.

Part b, 13 points

Find, by nodal analysis, a small-signal expression for V_{out}/V_{in} .

$V_{out}/V_{in} = \underline{\hspace{2cm}}$

~~AE~~ $I_I = 0$ @ V_{out}

6 $\left[-g_m \cdot v_{in} + (v_o - v_{in})/R_f = 0 \right]$

$v_o/R_f = v_{in} (g_m + 1/R_f)$

7 $\left[\frac{v_o}{v_{in}} = (1 + g_m R_f) \right]$

Part c, 5 points

$g_m = 10 \text{ mS}$ $R_f = 2000 \text{ Ohms}$
Give a numerical value for V_{out}/V_{in} .

$V_{out}/V_{in} = \underline{+ 21.}$

3.
$$\left[\frac{V_o}{V_i} = 1 + g_m R_f = 1 + 10 \text{ mS} \cdot 2000 \Omega \right.$$
$$= 1 + 20$$
$$= 21$$