

Prospect of Tunneling Green Transistor for 0.1V CMOS

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Abstract

Well designed tunneling green transistor may enable future VLSIs operating at 0.1V. Sub-60mV/decade characteristics have been clearly demonstrated on 8" wafers with statistical data. Large I_{ON} at low V_{DD} are possible according to TCAD simulations but awaits verification. V_{DD} scaling will greatly benefit from low (effective) band gap energy materials, which may be provided by type II heterojunctions of Si/Ge or compound semiconductors.

A Looming Barrier to IC Scaling

Reducing the voltage V_{DD} is a powerful way to reduce IC energy consumption, which is proportional to V_{DD}^2 . Power was kept under control because V_{dd} has been reduced in proportion to half-pitch up to 130nm as shown in Fig 1 [1]. The 14nm node is projected to operate at 0.7V, making the power consumption 25x larger than it would be if operated at 0.14V as the past trend suggests. While IC power consumption has been much discussed as a thermal management challenge, it also accounts for a few percent of the electricity demand and rising fast.

MOSFET current can not rise faster than one decade for every 60mV increase in V_g (the subthreshold swing). If one is to reduce the V_{DD} to 0.14V with I_{ON}/I_{OFF} ratio of mere 3 decades, SS needs to be 45mV/decade.

Green Transistor with Steep Turn On/Off

In MOSFETs (and BJTs), a potential barrier is raised and lowered by V_g to turn the current on and off (Fig. 2). Because of Boltzmann distribution, some electrons always have sufficient energy to pass over the barrier. The current can only be reduced by the rate of electron density drop with increasing energy, hence the 60 mV/decade limit. To beat this limit, the carriers must not flow over a barrier. They may tunnel through it.

Fig. 3 shows a prototype tunnel transistor [1-4]. Electrons are generated by band-to-band tunneling in the P+ source when the gate voltage bends the energy band to satisfy two conditions for tunneling—overlap of the valence and conduction bands and the presence of a high electric field or thin tunnel barrier[5]. The generated electrons flow through the surface N-channel to the drain. Fig. 4 shows simulated I_d - V_g [6] assuming uniform source doping. In the $1E\text{-}20\text{ cm}^{-3}$ case, very steep turn on occurs at the sudden onset of band overlap. Unfortunately the graded source diffusion (and the charge averaging effect of the Poisson equation) provide a continuous range of (effective) doping concentration near the tip of the source. At points of lower doping (Fig. 4) the turn-on voltages and electrical field (therefore I_{ON}) are both lower. The envelope of these curves is the IV of a prototype tunnel transistor, with a disappointing sub- V_t swing.

Fig. 5 shows a design that lowers the turn-on voltage of and thus enhances tunneling in the heavily doped part of the source [1]. The pocket doping has similar effect as the threshold implant in MOSFETs. Fig 6 shows the evolution of the IV of Si gFETs of this type as the pocket doping is increased [6]. Fig. 7 illustrates the concept of steep turn on at the sudden onset of the overlap of the conduction and valence bands.

Another approach is to use a very steep source doping profile to minimize the size of the low doping part of the source (but not the charge averaging effect). Fig. 8 shows that sub-60mV/decade average SS is achieved over six orders of magnitude of current with a device similar to Fig. 3 but fabricated on Si SOI substrate with a Ge source [7]. After gate and drain formation, the source region was etched and refilled with insitu doped poly-Ge at low temperature.

Fig. 9 shows yet another gFET design that produces steep turn on/off. The graded source region is minimized with dopant segregation from NiSi and a thin semiconductor pocket is created

between the gate dielectric and underlying NiSi [8]. The vertical electric field is enhanced in the thin pocket and further reduces the source edge effect. The measured 46 mV/decade SS could only be reproduced by simulation when a semiconductor pocket is present as shown in Fig. 9. Fig 10 shows the X-SEM images of the device with a wedge-shaped pocket between the silicide and the gate dielectric. This geometry is produced by recessing the source before the salicide steps. A control device without the pocket was also fabricated for comparison. Fig. 11 shows the measured I_D - V_G characteristics of the gFET vs. the control device. The SS of the gFET averages 46mV/dec over a decade of drain current. Another lot successfully reproduced the result two months later with 47mV/dec SS. As seen in Fig. 12, the gFET shows sub-60mV/dec SS over 3 decades of drain current. Special care must be taken to screen out low SS data due to various causes (Fig. 14). Statistical distributions of the minimum SS show that more than 30% of the gFETs on an 8" wafer have sub-60mV/dec SS (Fig. 15). The measured I_D - V_D characteristics of the gFET (Fig. 16) are similar to MOSFET's and typical for a tunnel transistor.

Prospects of Large I_{ON} at 0.1V V_{DD}

Simulations show that V_{DD} can be scaled by reducing E_g (Fig. 17). Only E_g was reduced in simulation, not the other tunneling parameters [1]. At $E_g=0.36\text{eV}$ I_{ON} exceeds $1\text{mA}/\mu\text{m}$ at $V_{DD}=0.2\text{V}$ with $CV/I=0.4\text{pS}$ and can operate well at 0.1V (Fig. 18). A type II hetero-junction gFET [9] provides an effective tunneling band gap (E_{geff}) lower than the E_g of both materials (Fig. 19, 20). Materials A and B can be simply strained Si thin film on Ge with theoretical E_{geff} of around 0.18eV (Fig. 21). A and B can be compound semiconductors such as InAs on AlGaSb with E_{geff} as small as zero by adjusting the Ga/Sb ratio (Fig. 22). The low density of states and strong quantization in require special attention in gFETs.

Conclusion

Carefully designed tunneling green transistors can potentially enable 0.1V microelectronics. Sub-60mV/dec subthreshold swing has been clearly demonstrated on 8" wafers and statistical data. High current and low voltage operation needs small tunneling band gap energy which may be provided by type II heterojunctions of compound semiconductors or Si/Ge. gFET in Fig. 5 is insensitive to gate length variations (Fig. 23) because the tunnel path is enclosed in the source and shielded from the influence of the drain.

Acknowledgement

Work partially supported by DARPA under a SPAWAR contract, #N66001-08-C-2022, FCRP/MSD, and NSF STC/E3S.

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Node (nm)	250	180	130	90	65	32	14
V_{dd} (V)	2.5	1.8	1.3	1.2	1.1	0.9	0.7

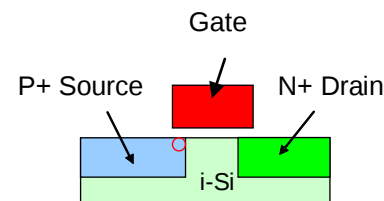
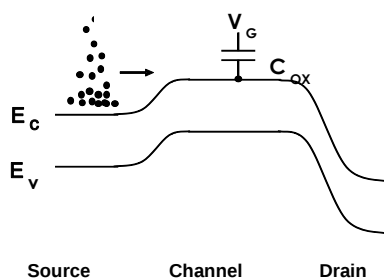


Fig. 1. IC V_{DD} scaling history and ITRS projection. Slowdown since 90nm leads to accelerated rise in energy consumption.

Fig. 2. Boltzmann statistics lead to the 60mV/decade limit because current is controlled with an energy barrier height.

Fig. 3. In a prototype N-type tunnel transistor, electrons are generated by tunneling in the red-circle region.

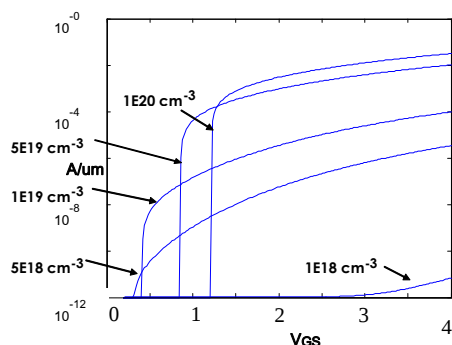


Fig. 4. I_{DS} is very sensitive to source doping level. Outer envelope of the curves leads to a poor sub- V_t swing of a tunneling transistor.

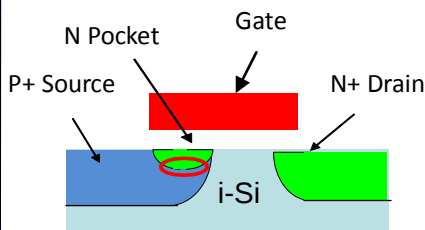


Fig. 5. One design of a low SS swing gFET. The P pocket lowers the turn-on voltage of the heavily doped source region.

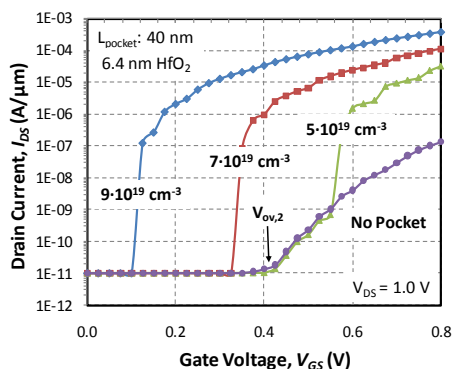


Fig. 6. Increasing N pocket doping makes the turn-on voltage of the 5E19cm-3 curve in Fig.4 lower than at edge of the source.

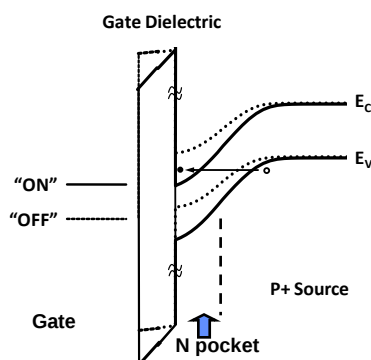


Fig. 7. Energy diagram from gate through pocket to source. Device is off when there is no overlap of valence and conduction bands.

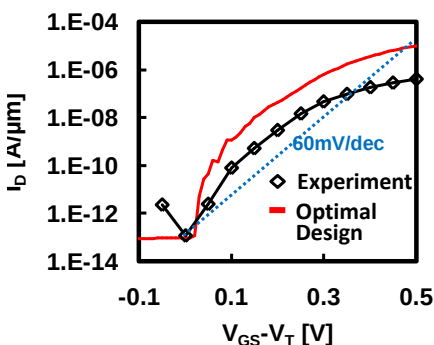


Fig. 8. Deposited insitu doped Ge source gFET produced sub-60mV/dec. average SS over 6 orders of current [7].

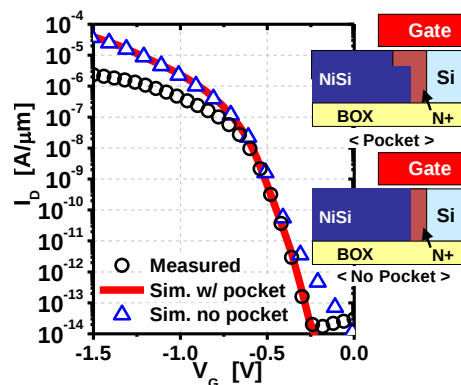


Fig. 9. Another design achieves <60mV/decade swing with a semiconductor pocket between oxide and metal (NiSi).

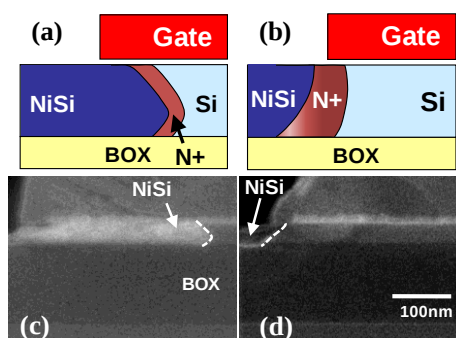


Fig. 10. Structure and cross-section SEM at source side of dopant segregated gFET (a), (c) and control device (b), (d).

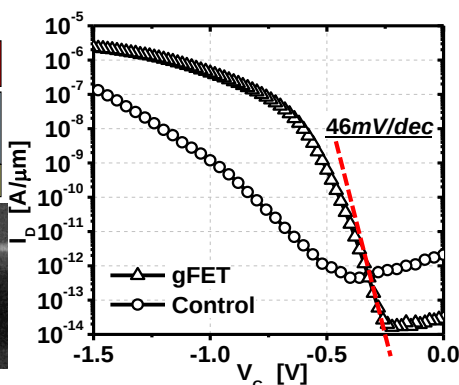


Fig. 11. I_D - V_G of measured and simulated gFET showing SS of 46mV/dec. ($L_G=20\mu m$, $V_{DS}=-1.0V$)

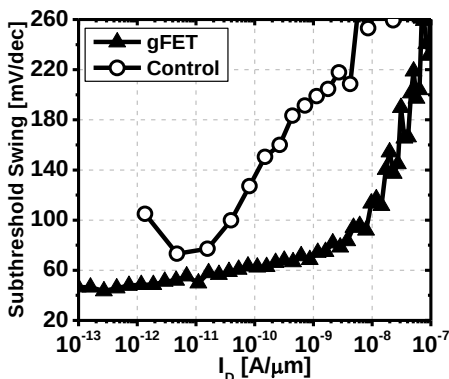


Fig. 12. SS < 60mV/dec over almost 3 decades of I_D for the gFET, but not seen in the control tunnel transistor.

	Ref. [2]	Ref. [3]	Ref. [4]	This Work
SS (mV/dec)	52.8	42	~300	46
I_{ON} ($\mu A/\mu m$)	12.1	0.01	1E-4	1.2
I_{ON}/I_{OFF}	1E4	1E4	1E2	7E7

Fig. 13. Comparison with other reported silicon tunnel transistors [8]. $V_{DS}=V_{GS}-V_{BTBT}=1.0V$.

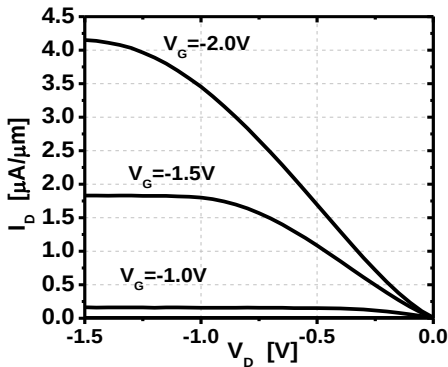


Fig. 16. I_D - V_D of gFET showing tunneling transistor behavior.

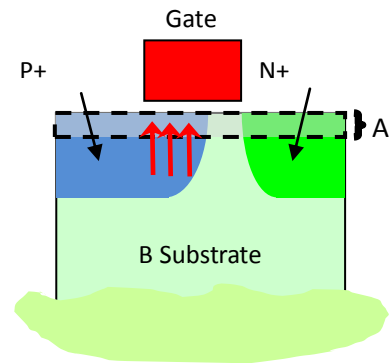


Fig. 19. A heterojunction gFET. See Fig. 20 for explanation.

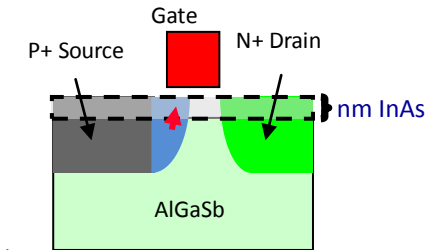


Fig. 22. gFET may employ heterojunction of compound semiconductor with tunable E_{eff} .

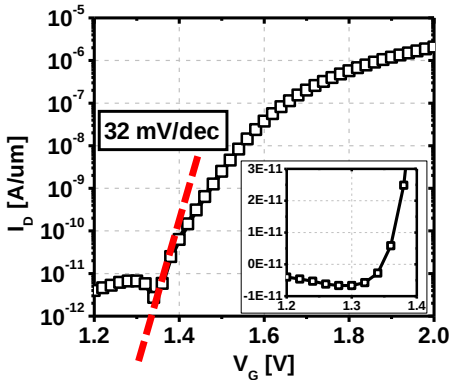


Fig. 14. Phantom low SS can show up without proper screening.

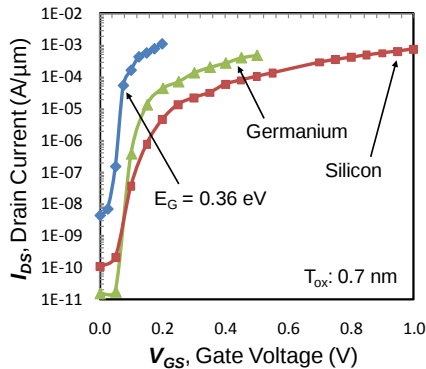


Fig. 17. Reducing the band gap energy is a new path for scaling the V_{DD} for gFET. CV/I is about or below 1 pS. Only E_g is varied.

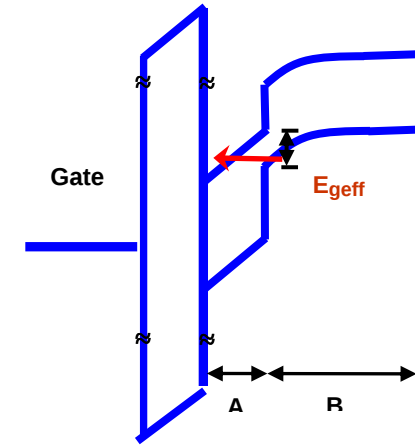


Fig. 20. The effective tunneling band gap in Type II heterojunction is smaller than those of A and B.

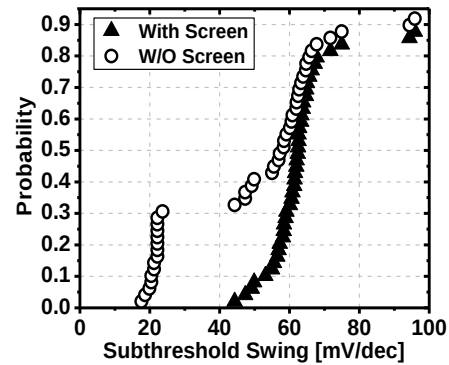


Fig. 15. SS distribution comparison of gFET and control device.

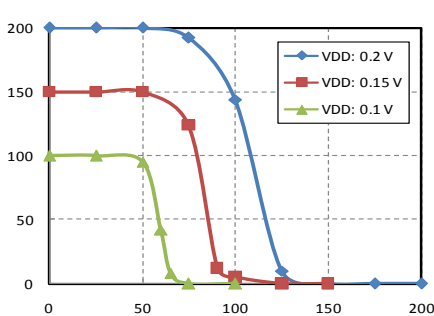


Fig. 18. P/N type gFETs of Fig. 17 with $E_g=0.36eV$ produce excellent inverter voltage transfer curves even at 0.1V.

Ref.	ΔEC	ΔEV	$E_{g,s-Si}$	$E_{g,eff}$
[10]	0.55	0.31	0.42	0.19
[11]	0.57	0.25	0.35	0.17
[12]	0.58	0.21	0.37	0.16

Fig. 21. Theoretical E_{eff} of a Ge-strained Si gFET is around 0.18eV, good for $V_{DD}<0.1V$ operation (see Fig. 17).

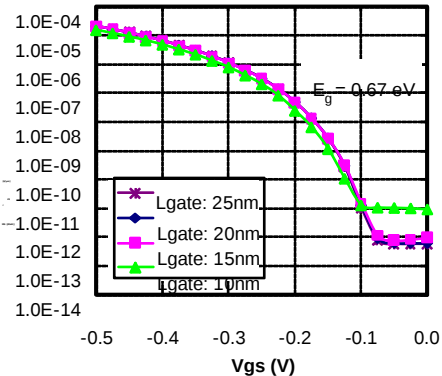


Fig. 23 gFET IV is insensitive to L_g variation because the tunneling region is in the N+ source, protected from the drain.